

Using VLSI Design Flow Outputs

EE241 Tutorial
Written by Brian Zimmer (2013)

1 Overview

In this tutorial, we will start with a fully place-and-routed 4-to-16 decoder created using the Synopsys VLSI design flow, import this design into Cadence Virtuoso, extract the design, and simulate it at the transistor level to verify predicted timing and energy results. The VLSI design flow uses a series of models and abstractions to generate circuits, so to truly understand the VLSI flow, it is instructive to remove these abstractions and compare the results in terms of timing and energy. Figure 1 shows a flowchart showing all of the conceptual steps in the tutorial.

2 Getting Started

2.1 Synopsys VLSI Flow

First, we will need to setup working directories for both the VLSI design flow. Run the commands below to setup the VLSI flow. On BWRC machines, replace `~ee241` with `/tools/designs/ee241` throughout this tutorial.

```
% source ~ee241/tutorials/ee241.bashrc
% cd /scratch/userA
% git clone ~ee241/tutorials/decoder_analysis
% cd decoder_analysis
% LABROOT=$PWD
```

This repository already contains a working 4-to-16 decoder, but you need to build the design yourself again. The following commands will run the entire flow to generate a place-and-routed design.

```
% cd $LABROOT/build-rvt
% make pt-pwr
```

Remember if you need to rerun any step in the process and the makefile claims that everything is up-to-date, you can run:

```
% make clean
```

2.2 Cadence Virtuoso

Next, run the commands below to setup a working directory for Cadence Virtuoso. Unlike the VLSI flow directory which will need to be copied for every different design, this directory only needs to be setup once and can be used for all of your projects.

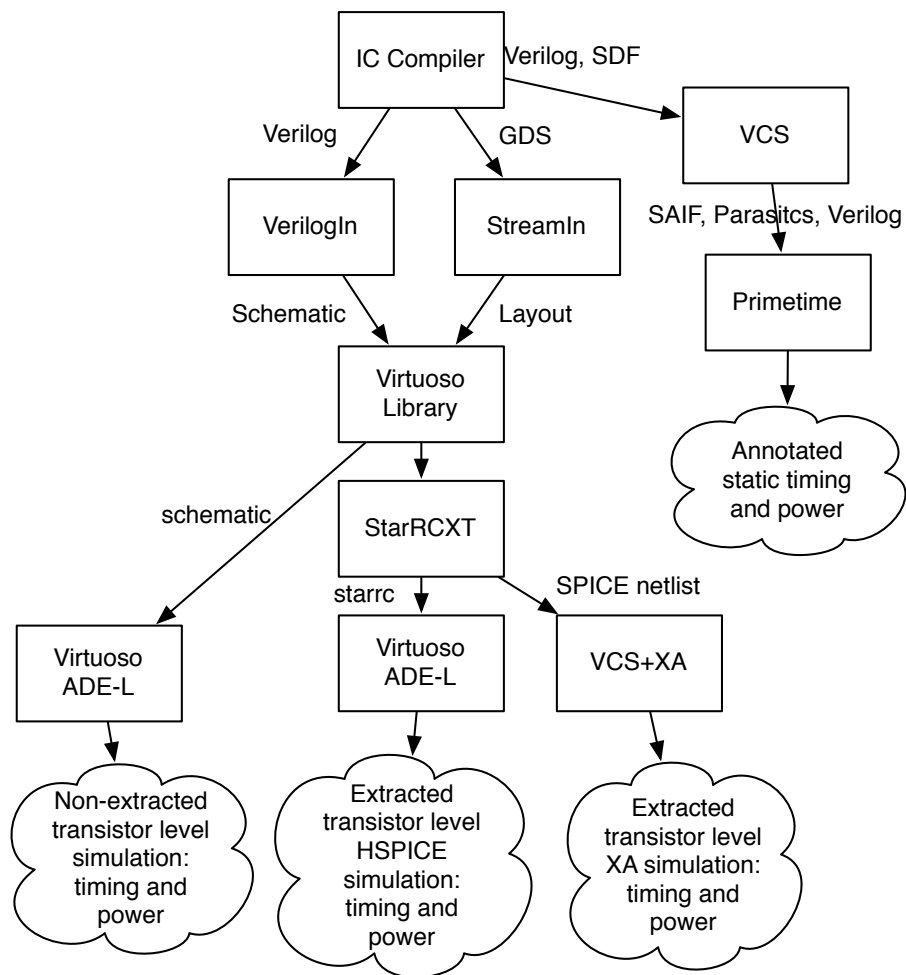


Figure 1: EE241 Toolflow for Lab 2

```
% cd /scratch/userA
% git clone ~ee241/ee241_virtuoso
% cd ee241_virtuoso
% VIRTUOSOLABROOT=$PWD
```

3 Translating layout

Open up the design in IC Compiler.

```
% cd $LABROOT/build-rvt/icc-par/current-icc
% ./start_gui
```

Notice that all that can be seen is the wiring—the standard cells are all hidden. To expose the design, on the left panel increase the level to “99” and click “Apply”, as shown in Figure 2.

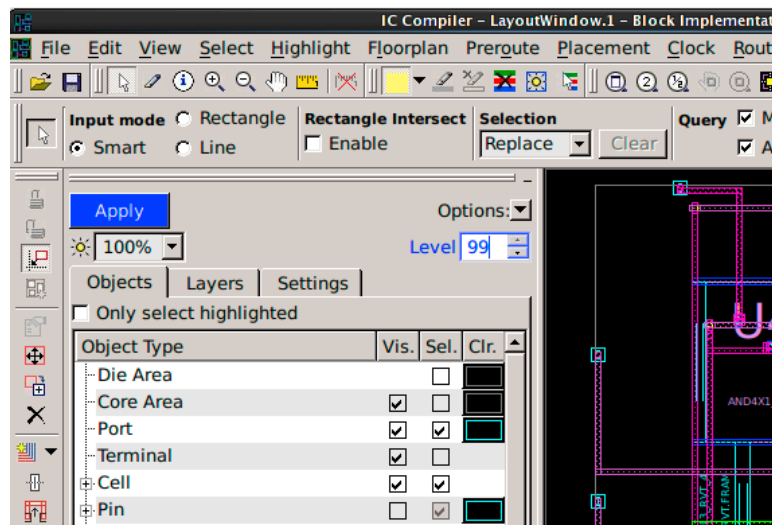


Figure 2: Exposing additional levels.

You can now see the metalization inside each standard cell, as shown in Figure 3 (use “z” and the mouse to zoom). But notice you still cannot see any devices—there are wells and poly but no active area. This will also be true of an IP blocks you place, such as memories. These views are only “black-box” views and only contain enough information to perform routing. To tape-out this design or perform transistor level extraction and simulation, we need all of the layers.

While IC Compiler is open, there are a few nice tools to use to better understand the output. For example, to highlight a path, go to Timing — New Timing Analysis Window. You can click “Apply” to see the worst paths, or set a specific path by setting the “From:” pin and “To:” pin. The TimingWindow will show the actual arrival time vs. the required arrival time, and report the difference as the “slack.” Negative slack will mean there is a violated setup time. Click on a path, and click “Inspector” to see the path as part of the schematic. Then, to highlight the path on the layout, click on Highlight — Inspected Path (or Selected Path).

Now, open Cadence Virtuoso.

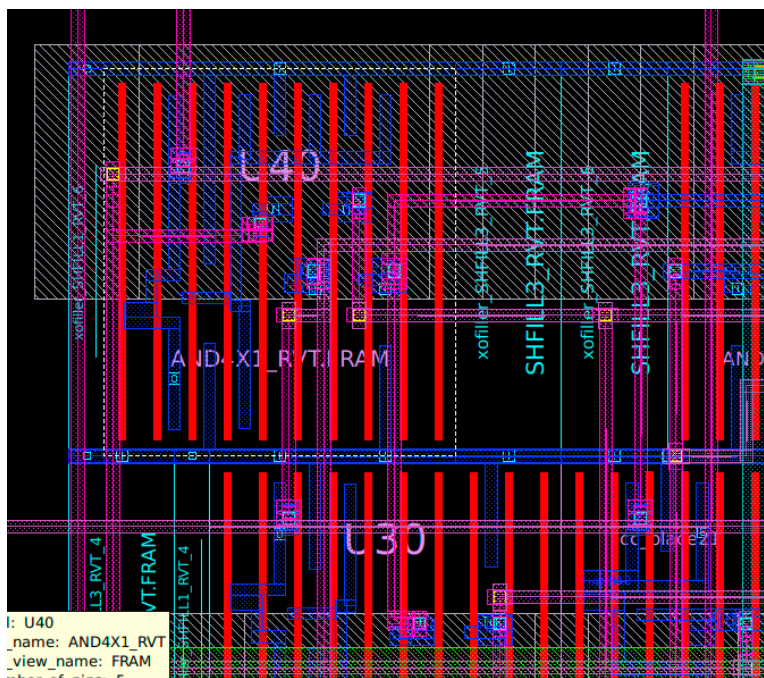


Figure 3: Blackbox view in IC Compiler.

```
% cd $VIRTUOSOLABROOT
% virtuoso &
```

Open the library `saed32nm_rvt`, then open the “layout” view of `AND2X1_RVT`. The layout is shown in Figure 4. Now press the `~` key to view only the frontend layers. The views available inside Virtuoso include all of the needed layers and are called “full-views.” You can cycle through other layers using the keys 1, 2, 3, etc to show that layer of metal and the via above it. Click “AV” to show all of the layers again.

We will use a GDS file format as an intermediate format to move from IC Compiler to Virtuoso. The GDS file will include instantiations of every standard cell, and we will intercept these instantiations and point them to use the “full-views” available in Virtuoso. The GDS file has already been generated inside `icc-par/current-icc/results/decoder.gds`.

In Virtuoso, make a new library by going to File — New — Library. Give it a name, then select “Attach to an existing technology library” and choose `SAED_PDK_32_28`.

Then go to File — Import Stream, then enter:

- File — Import — Stream
- “Stream File”: `icc-par/current-icc/results/decoder.gds`
- “Library”: The library you just created
- “Top Level Cell”: `change_names_icc`
- “Attach Technolog Library”: `SAED_PDK_32_28`
- Then click “Show Options”

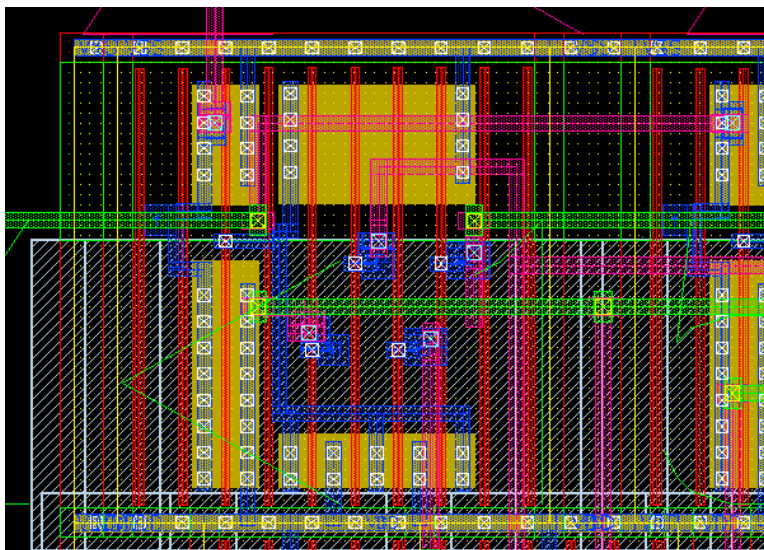


Figure 4: Full view in Virtuoso.

- Check the box next to "Replace [] with <>" (Note: if you do not see this option, open the layout after streamin and replace all pin label text square brackets with angle brackets)
- Click the "layers" tab.
- Under "map file", click open, and open ee241/synopsys-32nm/techfiles/saed32nm_1p9m_gdsout.map.bck
- Click on the "libraries" tab
- Click on saed32nm_rvt, then click on the right arrow
- Click the "cells" tab
- Click "Add row"
- Enter: Stream name: "change_names_icc", library "test" (or the name of the library you created), "cell": decoder, view: "layout"

Click apply, then open the layout view of the design in your new library. Press Shift-f to expose the standard cells so you can see the complete design, as shown in Figure 5.

4 Translating schematics

In order to run extraction or LVS on this design, we need to generate schematics for this design.

Note that this design started as RTL level verilog. This is useless as a schematic, as there are no standard cells yet. We could use synthesis output because this includes structural Verilog, yet there is no clock tree and ICC can change cells in order to fix timing, so it will not match the layout. Therefore we need to use the structural Verilog generated by IC Compiler. Open this output in your favorite editor with the following commands

```
% cd $LABROOT/build-rvt/icc-par/current-icc/results
% vim decoder.output.v
```

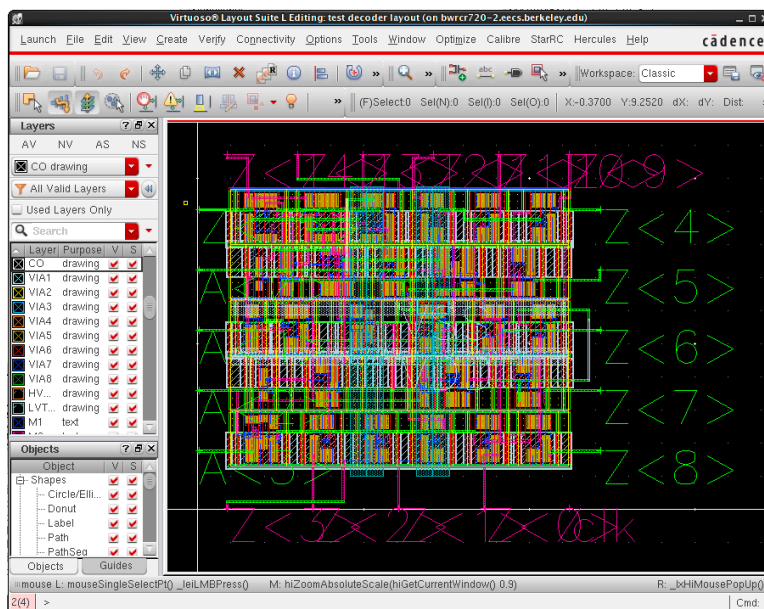


Figure 5: Entire decoder imported into Virtuoso.

Notice that each standard cell does not include power and ground. Power and ground was not in the input netlist, and power is treated specially in IC Compiler.

Go back to Virtuoso, and open the schematic and symbol views of the AND2X1_RVT cell in saed32nm_rvt. Notice that this view expects VDD and VSS, because a transistor level netlist needs to know what net is connected to the transistor sources. Therefore our Verilog description needs to include information about power as well. There is another file in the output of IC Compiler that includes this information.

```
% cd $LABROOT/build/icc-par/current-icc/results
% vim decoder.pg.lvs.v
```

Notice that this is still a Verilog netlist. Just like the layout from IC Compiler has no information about the actual devices, this Verilog netlist has no information about transistors. However, Virtuoso's schematics are a “full-view” that include the devices, so during import to Virtuoso, we will need to reference the standard cell library as we did for the layout import.

Go to File — Import — Verilog, then and load “verilogin_template”. The form should look like Figure 6.

You will need to change:

- Target Library Name: your new library name
- Verilog file to import: icc-par/current-icc/results/decoder.output.pg.lvs.v
- -v Options: ~ee241/stdcells/synopsys-32nm/vendor/lib/stdcell_rvt/verilog/saed32nm.all_pins.v

Click apply, then open the generated schematic. Virtuoso has instantiated all of the standard cells and routed all of the nets.

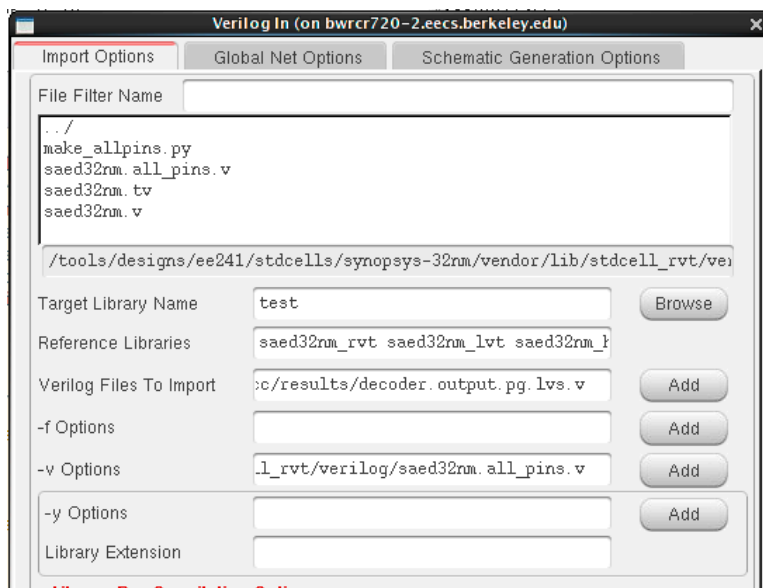


Figure 6: VerilogIn dialog settings.

5 LVS

In order to perform parasitic extraction of a design, the design must pass layout-versus-schematic (LVS). This is because the parasitics caused by polygons in layout need to be annotated to their equivalent net in the schematic.

To run LVS, open the layout view and go to Hercules — Run LVS. Wait a few seconds for the “Block” entry to be filled in automatically as shown in Figure 7, then press “Execute.” You should see LVS pass as shown in Figure 8. Once you have verified that the design has been imported correctly (and ICC has done its job correctly), you can run extraction.

6 Extraction

To run parasitic extraction in Virtuoso, open the layout view and go to StarRC — Parasitic Generation Cockpit. The setting should be automatically loaded for you as shown in Figure 9, but you will need to change one thing: Click on the “Output Parasitics” tab, and check the box next to “Ports Annotation”, then enter in your library and cell name. If you do not do this, you will not be able to netlist the design. Some other options are useful to understand.

- Extract Parasitics Tab — Extraction. RC will extract both resistance and capacitance.
- Extract Parasitics Tab — Couple to Ground = NO will include coupling between nets in your extraction
- Extract Parasitics Tab — Additional Options, allows you to set a threshold to try to reduce device count

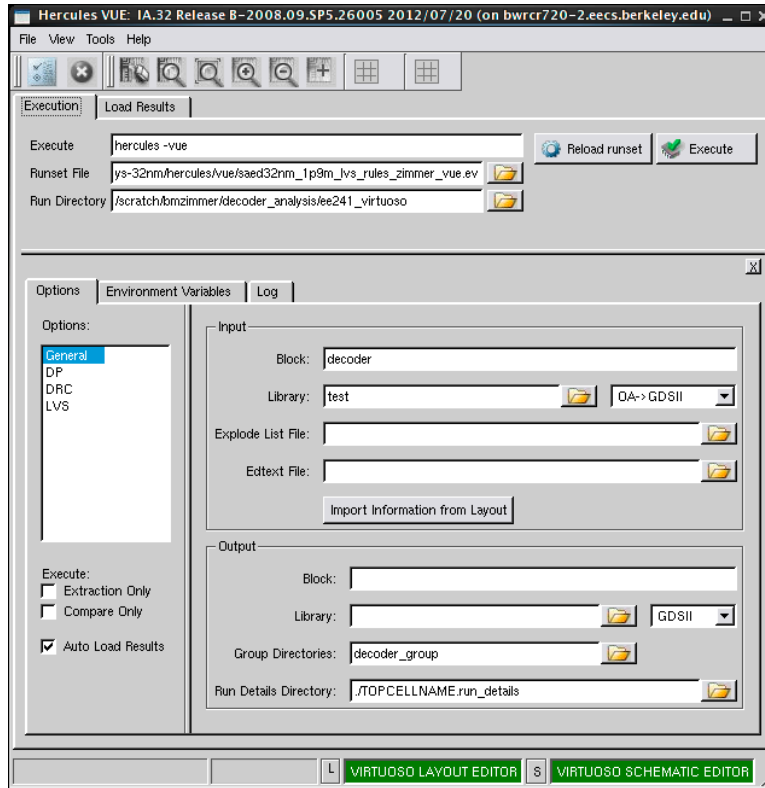


Figure 7: Hercules LVS dialog settings.

TOP BLOCK COMPARE RESULTS
CLEAN

[decoder == decoder]

Hercules (R) Hierarchical Design Verification, IA.32 Release B-2008.09.SP5.26005 2012/07/14
Hercules Run: Time= 0:00:01 User=0.40 System=0.21

Netlist Extraction Statistics

Library name: /scratch/fmzimmer/decoder_analysis/ee241_virtuoso/decoder.gdsOut
Structure name: decoder

Extraction Errors:

Layout vs. Schematic Statistics

Schematic: /scratch/fmzimmer/decoder_analysis/ee241_virtuoso/decoder.sch_out
Comparison completed with 1 successful equivalencies
Comparison completed with 1 failed equivalencies

Figure 8: Decoder passing LVS.

Now click “Apply” and wait a minute while extraction runs. When it finishes, go to your new library and open the “starrrc” view of the decoder cell. If you zoom in, you can see the annotated parasitics overlaid on your layout as shown in Figure 10. By going to StarRC — Parasitic Prober you can explore the parasitics in your design.

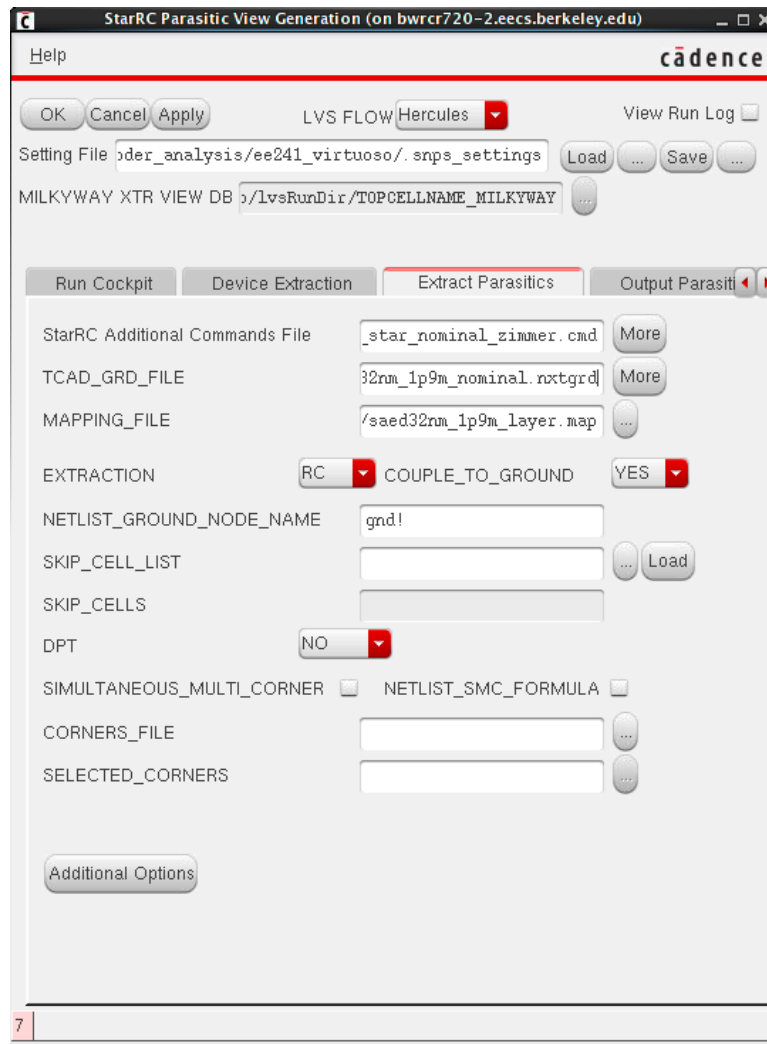


Figure 9: StarRC settings.

7 Simulation in Virtuoso

Go to File — New — Cell View, call it decoder.tb, and make the view and type “schematic”. Press ‘i’, then insert the symbol view of your newly imported decoder. Using the “vdc” and “vpwl” components in analogLib and the “vdd”, “gnd”, “noConn” components in basic, hook up VDD, VSS, and the inputs as shown in Figure 11. Note that you need to instantiate an array of noConn by pressing q while the noConn block is selected and changing the instance name, as shown in Figure 12. Set the VPWL as follows:

- A[3]: 3 pairs of points, T1: 0, V1: 1.05, T2: 500p, V2: 1.05, T3: 510p, V3: 1.05
- A[2]: 3 pairs of points, T1: 0, V1: 0, T2: 500p, V2: 0, T3: 510p, V3: 1.05
- A[1]: 3 pairs of points, T1: 0, V1: 1.05, T2: 500p, V2: 1.05, T3: 510p, V3: 1.05
- A[0]: 3 pairs of points, T1: 0, V1: 1.05, T2: 500p, V2: 1.05, T3: 510p, V3: 1.05

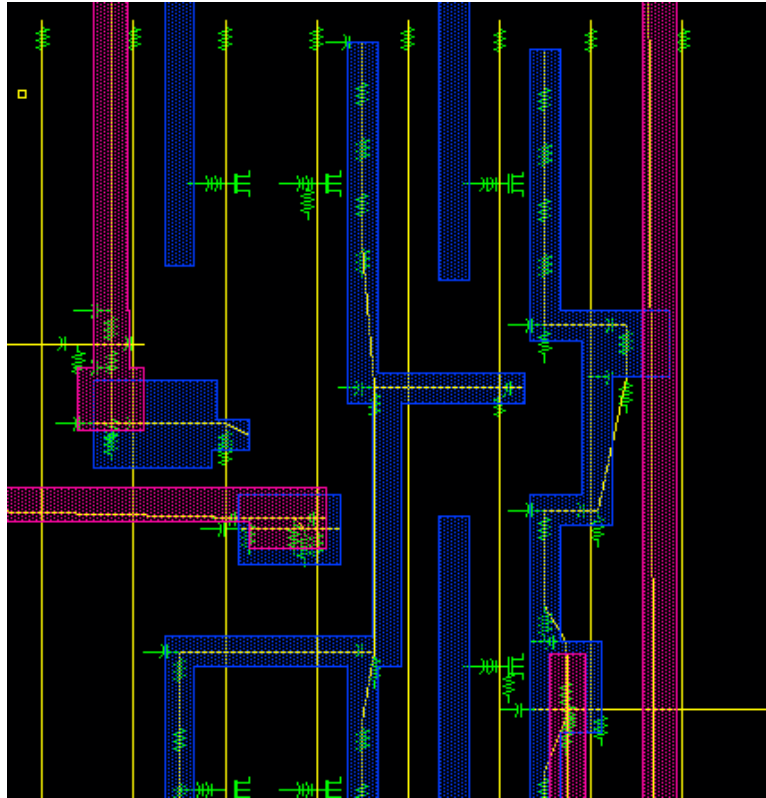


Figure 10: Extracted resistors and capacitors annotated on layout.

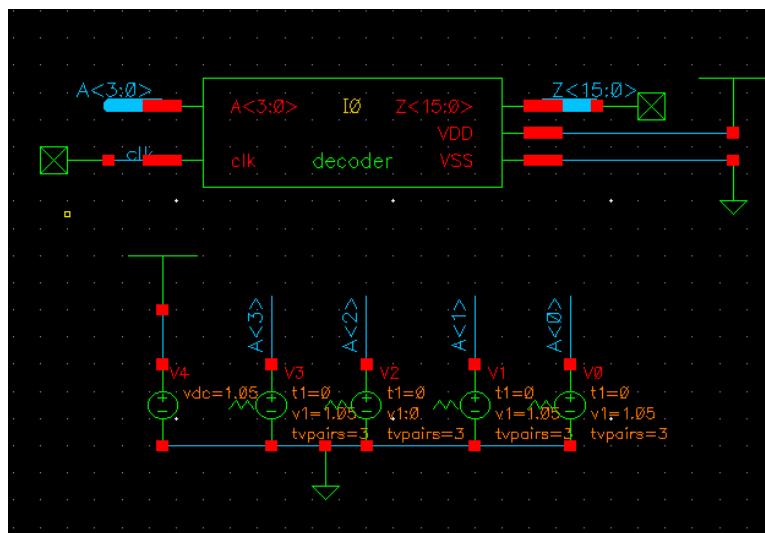


Figure 11: Testbench to measure a timing path.

Go to File — Check and Save, and there should be no errors or warnings. Next, we need to make a “config” view that will let us tell the simulator whether to simulate with the schematic view or the starrc view. Go to File — New — Cellview, and enter

Property	Value	Display
Library Name	basic	off
Cell Name	noConn	off
View Name	symbol	off
Instance Name	I4<15:0>	off

User Property	Master Value	Local Value	Display
nlAction	ignore		off

Buttons: OK, Cancel, Apply, Defaults, Previous, Next, Help

Figure 12: Creating an array of noConn instances.

- Cell: decoder_tb (the testbench you just created)
- View: config

Then click enter the settings shown in Figure 13. When the next window opens, go to File — Save. Now in the schematic window, go to Launch — ADE L. Go to Setup — Simulator, and make sure HSPICE is chosen. Click on Setup — Design, and choose View: config, and press ok. Next, go to Analysis — Choose, and set the stop time to 1n, and click ok. Then go to Outputs — To be Plotted — Select on Schematic, and click on both the A and Z bus (and select all of their signals), then press “Esc” when finished. Last, go to Simulation — Netlist and Run, and you should see a result like Figure 14.

To measure the path we are measuring (A[2] rising to Z[15] rising), mouse over the first edge at the 50% point and press “a”, then go to the second edge and press “b”.

Your result is only measuring the devices. Now, we would like to simulate this same path, but also annotate the parasitics we calculated during extraction.

Open the config view of decoder_tb, (click “yes” for Configuration config), then in the “View to Use” column for the “decoder” instance, enter starrrc as shown in Figure 15. Then go to File — Save.

Go back to ADE L and rerun the simulation. The delay should now increase because the parasitics of the wires are now included. Go to File — Save State, then save inside “cellview” so you can open this simulation later. By going to Simulation — Netlist — Display, you can make sure the parasitics are being included within the spice deck.

Note that IC Compiler and Primetime perform timing on all paths. However, many paths in the decoder are false paths. For example, the path A[3] falling to Z[2] falling is a false path. Because it is a one hot decoder, a high A[3] means that only Z[15:8] could be high, so there is no possible way for Z[2] to fall.



Figure 13: Settings for the new config view.

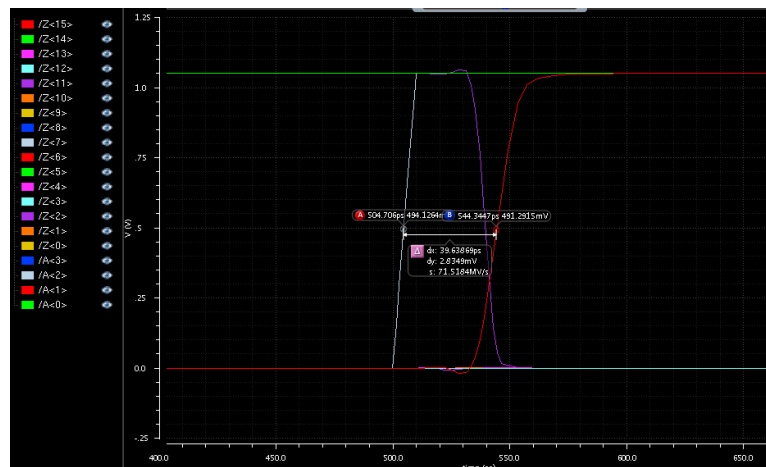


Figure 14: A[2] rise to Z[15] rising timing path for schematic simulation.

Library	Cell	View Found	View To Use	Inherited View List
SAED_PDK_32_28	pmos4t	hspice		hspiceD hspice ...
analogLib	vdc	hspiceD		hspiceD hspice ...
analogLib	vpulse	hspiceD		hspiceD hspice ...
saed32nm_rvt	AND2X1_RVT	schematic		hspiceD hspice ...
saed32nm_rvt	AND4X1_RVT	schematic		hspiceD hspice ...
saed32nm_rvt	INVX0_RVT	schematic		hspiceD hspice ...
saed32nm_rvt	INVX1_RVT	schematic		hspiceD hspice ...
saed32nm_rvt	INVX2_RVT	schematic		hspiceD hspice ...
saed32nm_rvt	NBUFFX2_RVT	schematic		hspiceD hspice ...
saed32nm_rvt	NOR2X0_RVT	schematic		hspiceD hspice ...
saed32nm_rvt	NOR2X2_RVT	schematic		hspiceD hspice ...
saed32nm_rvt	SHFILL1_RVT	schematic		hspiceD hspice ...
saed32nm_rvt	SHFILL2_RVT	schematic		hspiceD hspice ...
saed32nm_rvt	SHFILL3_RVT	schematic		hspiceD hspice ...
test	decoder	starrc	starrc	hspiceD hspice ...
test	decoder_tb	schematic		hspiceD hspice ...

Figure 15: Change the config view to use the extracted netlist.

7.1 Energy measurement

At the beginning of the tutorial when you typed “make pt-pwr”, you should simulated the post-place-and-route netlist with annotated parasitics and performed power analysis on your design. Go back to the simulation to see what activity the testbench generated.

```
% cd $LABROOT/build-rvt/vcs-sim-gl-par
% make convert
% dve -vpd vcdplus.vpd &
```

Then plot both the A and Z signals. You should see the decoder is driven with A counting from 0000 to 1111 as shown in Figure 16. By running “make convert”, the switching activity of every node is annotated in a SAIF file. Then Primetime (in pt-pwr) combines this with the icc-par netlist and parasitics file (SBPF) to accurately measure energy consumption. For example, look at the generated SAIF file (switching activity).

```
% cd $LABROOT/build-rvt/vcs-sim-gl-par
% make convert
% cat vcdplus.saif
(INSTANCE U34
  (NET
    (A1
      (T0 1600) (T1 1400) (TX 0)
      (TC 1) (IG 0)
    )
    (A2
      (T0 1400) (T1 1590) (TX 10)
      (TC 14) (IG 0)
    )
    . . . .
```

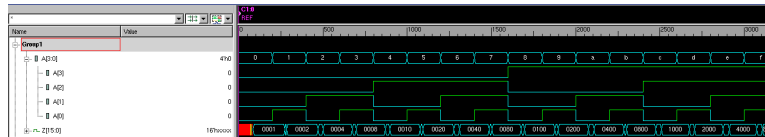


Figure 16: Test vector generated by Verilog testbench.

TC counts the number of toggles, and is used to calculate dynamic energy. T1, T0, and TX correspond to time at logic 1, 0, and X and are used to calculate leakage energy (leakage will be state dependent).

Now look at the Primetime energy report:

```
% cd $LABROOT/build-rvt/pt-pwr
% make
% cat current-pt/reports/vcdplus.timing.avg.max.report
```

Hierarchy	Switch Power	Int Power	Leak Power	Total Power	%
decoder	3.09e-05	3.41e-05	2.22e-06	6.72e-05	100.0

So for the 3.2ns testbench, the average power was 67uW for the decoder.

Now, back in Virtuoso, recreate the same testbench inside decoder.tb as shown in Figure 17 using vpulse with periods of 400ps, 800ps, 1600ps, and 3200ps for A[0], A[1], A[2], and A[3], and pulse time as half of the period. You might want to backup your old testbench. Launch ADE again, and run a transient simulation for 3.2ns. Probe the current through the VDD current source. Then go to Outputs — Setup, click New Expression, and set the Name to be “power”, then next to calculator click Open. Click the “it” button, then select the positive port of the VDD voltage source. Then in the function panel, click “Average”, then append “1.05” to the expression in the textbox. Without closing the calculator, go back to the Setup Output Options dialog and click “Get Expression”. The box will look like Figure 18. Click “Add”, then “Ok”. Now run the simulation. In the outputs pane, you should see a number for your power. This is the power measured by a spice simulation of the extracted netlist.

8 Mixed-mode simulation

For small designs, full HSPICE simulation in Virtuoso using handcrafted testbenches is easy. However, for larger designs, it is very useful to stimulate the simulation with a Verilog testbench, using a faster simulator, such as XA. Then critical analog components such as PLLs, SRAMS, and asynchronous interfaces can be simulated at the transistor level while the basic digital logic can be simulated more quickly with VCS.

Another reason to run mixed-mode simulations is to measure the effects of voltage scaling on energy. While Primetime works well for measuring power of digital logic, voltage scaling studies require

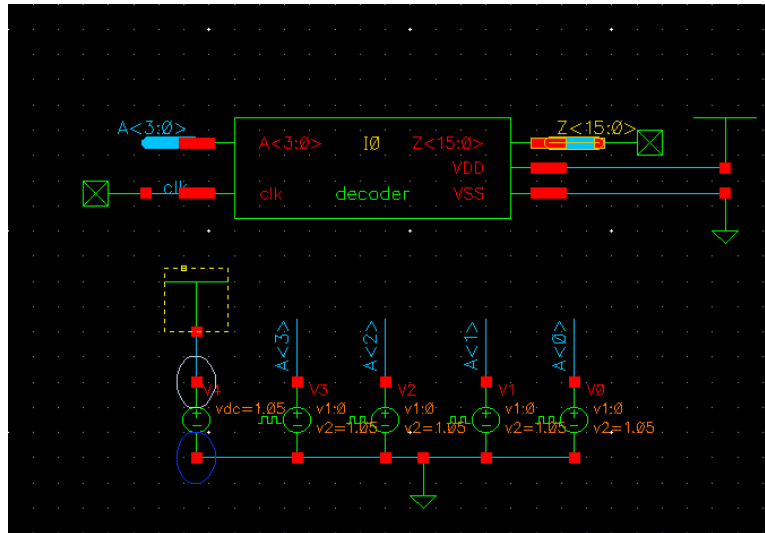


Figure 17: Test bench inside Virtuoso that emulates the Verilog testbench.

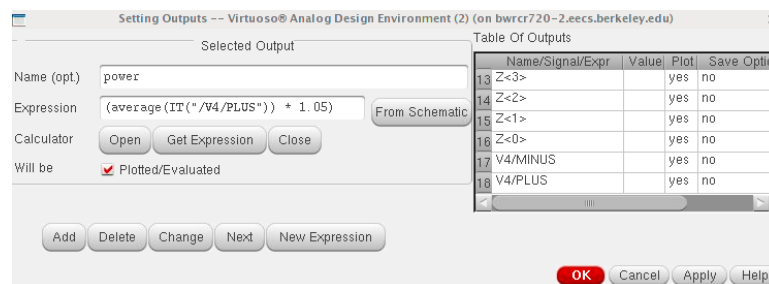


Figure 18: Create a measure statement to calculate average power.

all of the standard cells to be characterized at different voltages, and these cells are not always available.

In ADE L, go to Simulation — Netlist — Create. A netlist with parasitics should appear. Now to go File — Save As, and place this design in the \$LABROOT/build-rvt/vcs-sim-tl-pex directory as decoder.sp. This directory is a new directory that will run a mixed simulation.

Look at the Makefile in this directory.

```
% cd $LABROOT/build-rvt/vcs-sim-tl-pex
% cat Makefile
```

```
srcdir = $(basedir)/src
vsrsrc = \
    $(srcdir)/decoder_tb.v
```

...

```
$(vcs_sim) : Makefile $(vsracs)
    $(VCS) $(VCS_OPTS) +incdir+$(srcdir) -o $(vcs_sim) \
        +define+CLOCK_PERIOD=$(vcs_clock_period) \
        $(vsracs)

...

VCS_OPTS = -notice -PP -line +lint=all,noVCDE,noTFIPC,noIWU,noOUDPE \
    +v2k -timescale=1ns/1ps \
    -P ../icc-par/current-icc/access.tab -debug \
    +neg_tchk \
    -ad=vcsAD.init -lca
```

The `-ad=vcsAD.init` line is the magic line that enables mixed signal simulation. Think of the `vcsAD.init` file as the config file we used in Virtuoso to choose between schematic only and parasitic simulation. You can replace `$(vcs_clock_period)` to define a clock period in terms of ns.

The only Verilog source is the same testbench as before, which instantiates the decoder with:

```
decoder DUT0 (.A(A), .Z(Z), .clk(clk));
```

Now look at the configuration file. This defines the analog to digital interface, tells VCS to use the Synopsys XA simulator (a transistor level simulator like HSPICE but optimized for speed), and to use spice on the “decoder” cell corresponding to the instantiation above. Also, Verilog has bus format for signals (eg. `A[3:0]`) but SPICE has no buses, so the tool must know how to convert between formats to hook up the ports correctly. `xa_options` tells the simulator to merge all probed signals with the VPD file so you can see digital and analog signals together in the same tool.

```
% cat vcsAD.init
choose xa -hspice decoder.sp -c xa_options -o xa;
use_spice -cell decoder;
bus_format <%d>;
d2a rf_time=10p rise_time=20p fall_time=20p delay=0 hiv=1.05 lov=0.0 node=*;
a2d loth=0.3v hith=0.7v node=*;
```

Next we need to hookup the supplies within the decoder. While there are cleaner ways to do this, we will do something easy for now. Edit `decoder.sp`. Delete all lines (except the first line) before the `.subckt` definition, and all lines after the `.ends` at the end of the circuit. *Make sure the first line is a comment.* Note that if your power measurement fails, you need to make sure your FROM and TO values are within the simulation time. Then add a path to the library, and place voltage sources as shown below:

```
% cat decoder.sp

* Decoder
.lib /tools/designs/ee241/synopsys-32nm/hspice/saed32nm.lib TT

.subckt decoder a<3> a<2> a<1> a<0> vdd vss z<15> z<14> z<13> z<12> z<11>
```



```
+ z<10> z<9> z<8> z<7> z<6> z<5> z<4> z<3> z<2> z<1> z<0> clk
Vvdd vdd 0 DC=1.05
Vvss vss 0 DC=0
.probe tran i(Vvdd)
.measure pavg AVG i(Vvdd) FROM=0 TO=3.2n

cg1431 u26|mp3|drn 0 c=85.1142e-18
```

Last, run the simulation and see the power measured by XA.

```
% make
% make run
% cat xa.meas
% dve -vpd vcdplus.vpd
```

9 Conclusion

In this tutorial, you investigated the accuracy of IC Compiler and Primetime reports. In order to run a fully-extracted transistor-level simulation, you need to import both the GDS and Verilog netlist generated by IC Compiler into a Virtuoso schematic and layout. Then using Hercules you run LVS to verify the import, and using Synopsys StarRCXT you extract parasitics from the layout. With this information, you can simulate a path and compare its delay between schematic-only simulation, extracted simulation, and IC Compiler estimates. Also, you can measure energy based on extracted simulation results, primetime, and a mixed signal simulation of an extracted netlist through a Verilog testbench.